

Digital Communication Systems

ECS 452

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6. Digital Modulation



Office Hours:

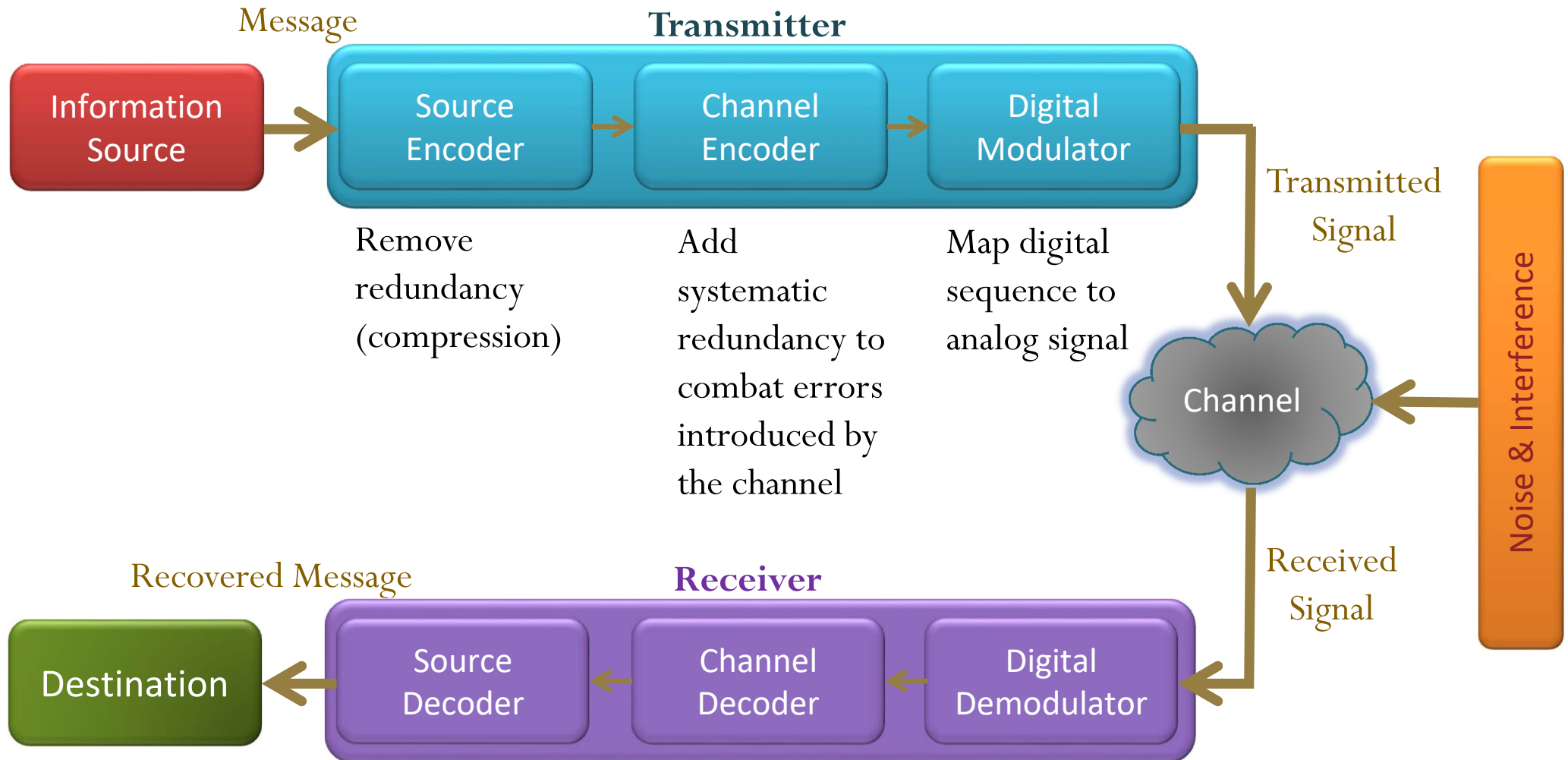
BKD, 6th floor of Sirindhralai building

Monday 10:00-10:40

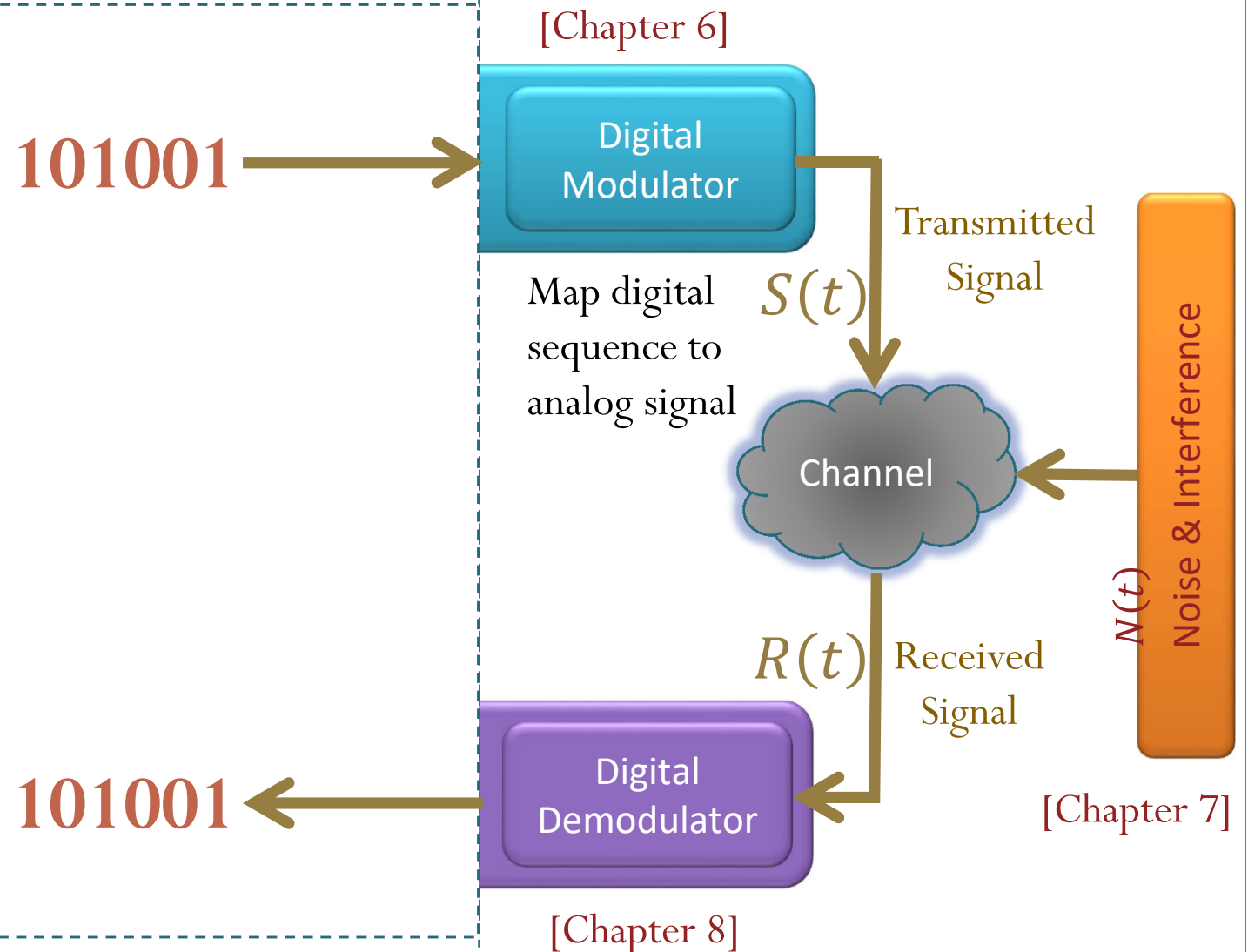
Tuesday 12:00-12:40

Thursday 14:20-15:30

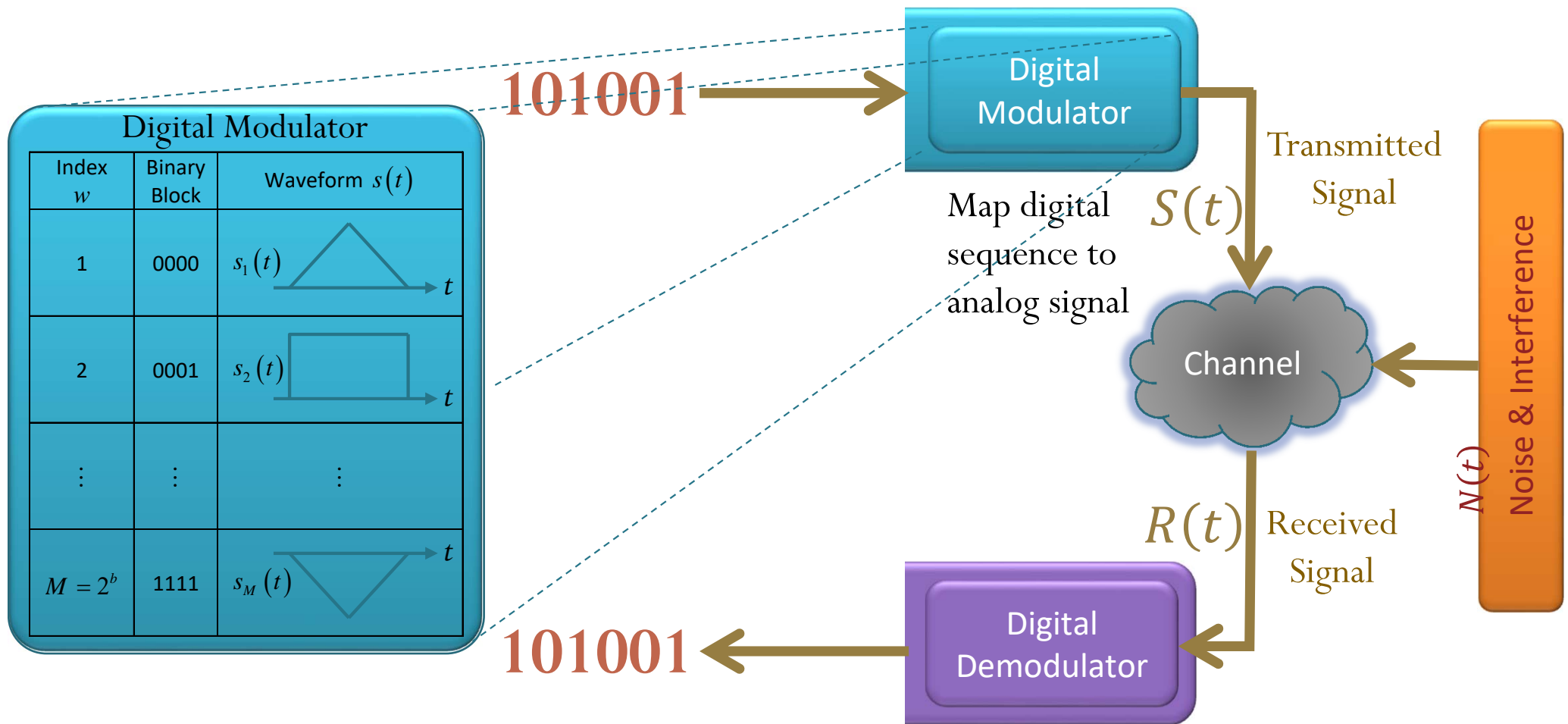
Elements of digital commu. sys.



Digital Modulation/Demodulation



Digital Modulation/Demodulation



Energy and Power

- Consider a signal $g(t)$.
- Total (normalized) **energy**:

$$E_g = \int_{-\infty}^{\infty} |g(t)|^2 dt = \lim_{T \rightarrow \infty} \int_{-T}^T |g(t)|^2 dt \stackrel{\text{Parseval's Theorem}}{=} \int_{-\infty}^{\infty} |G(f)|^2 df.$$

$$\Psi_g(f) = |G(f)|^2$$

ESD: Energy Spectral Density

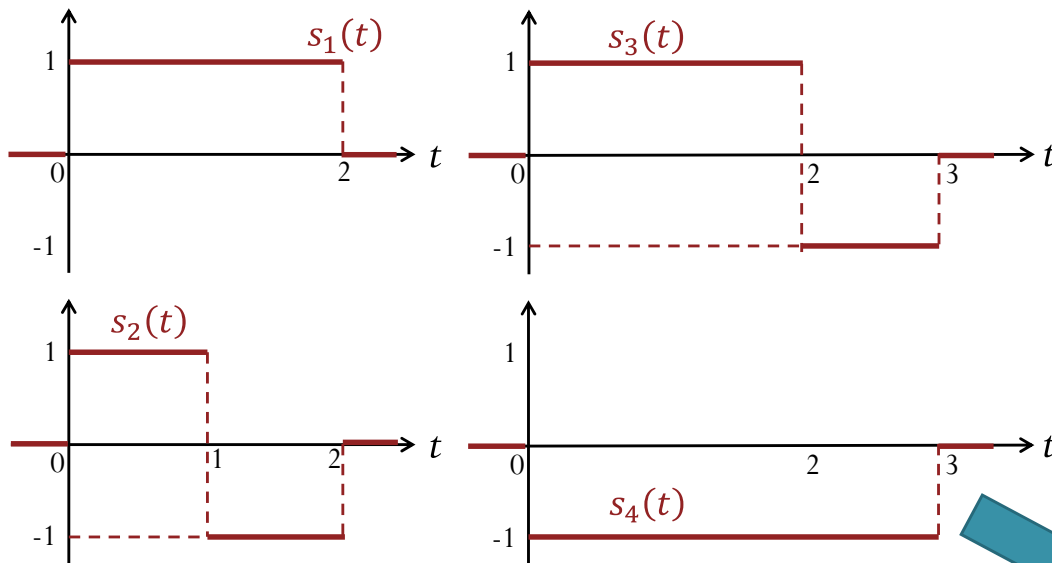
- Average (normalized) **power**:

$$P_g = \left\langle |g(t)|^2 \right\rangle = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} |g(t)|^2 dt = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |g(t)|^2 dt.$$

time-average operator

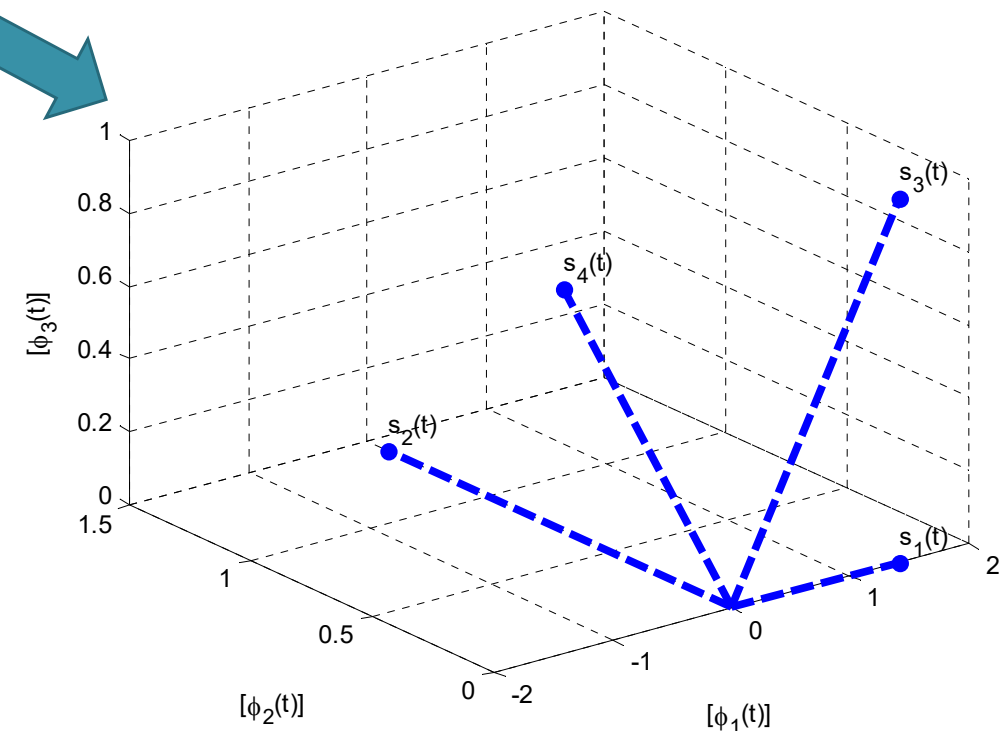


Review: From Waveforms to Constellation



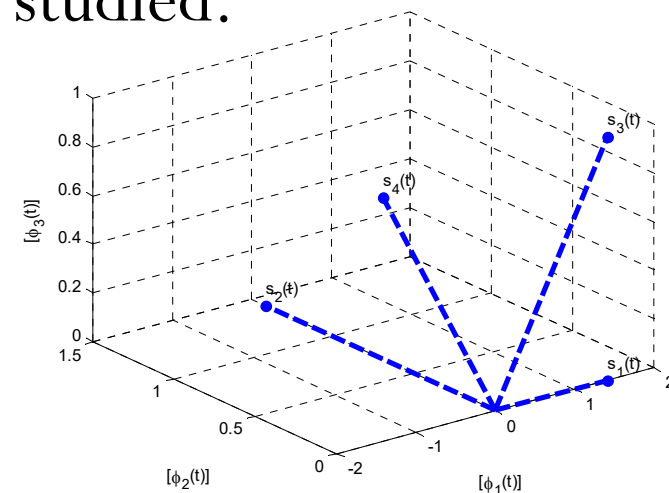
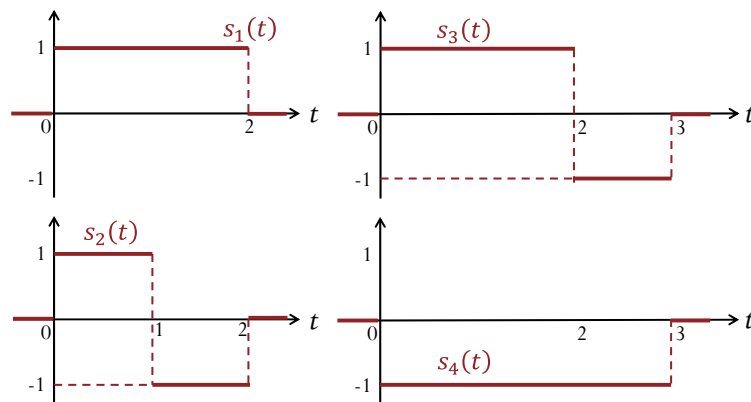
- What should be the axes?
 - 1 axis = 1 dimension
 - How many do we need?
- How to represent the waveform on these axes?

- A waveform contains infinitely many points. To represent *all* possible waveforms, we would need to work in infinite-dimensional space.
- However, we only have to consider four possible waveforms here.

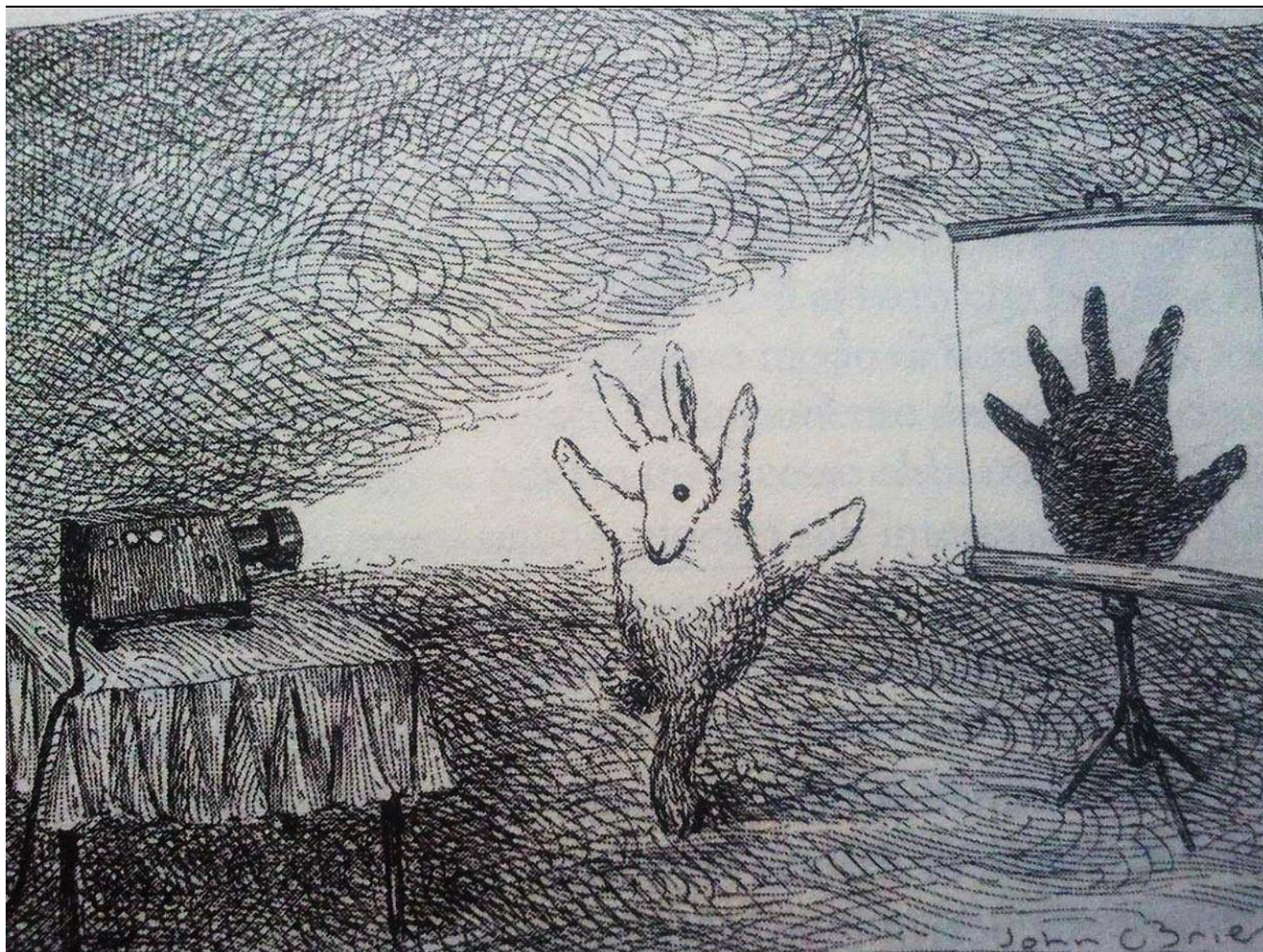


Review: From Waveforms to Constellation

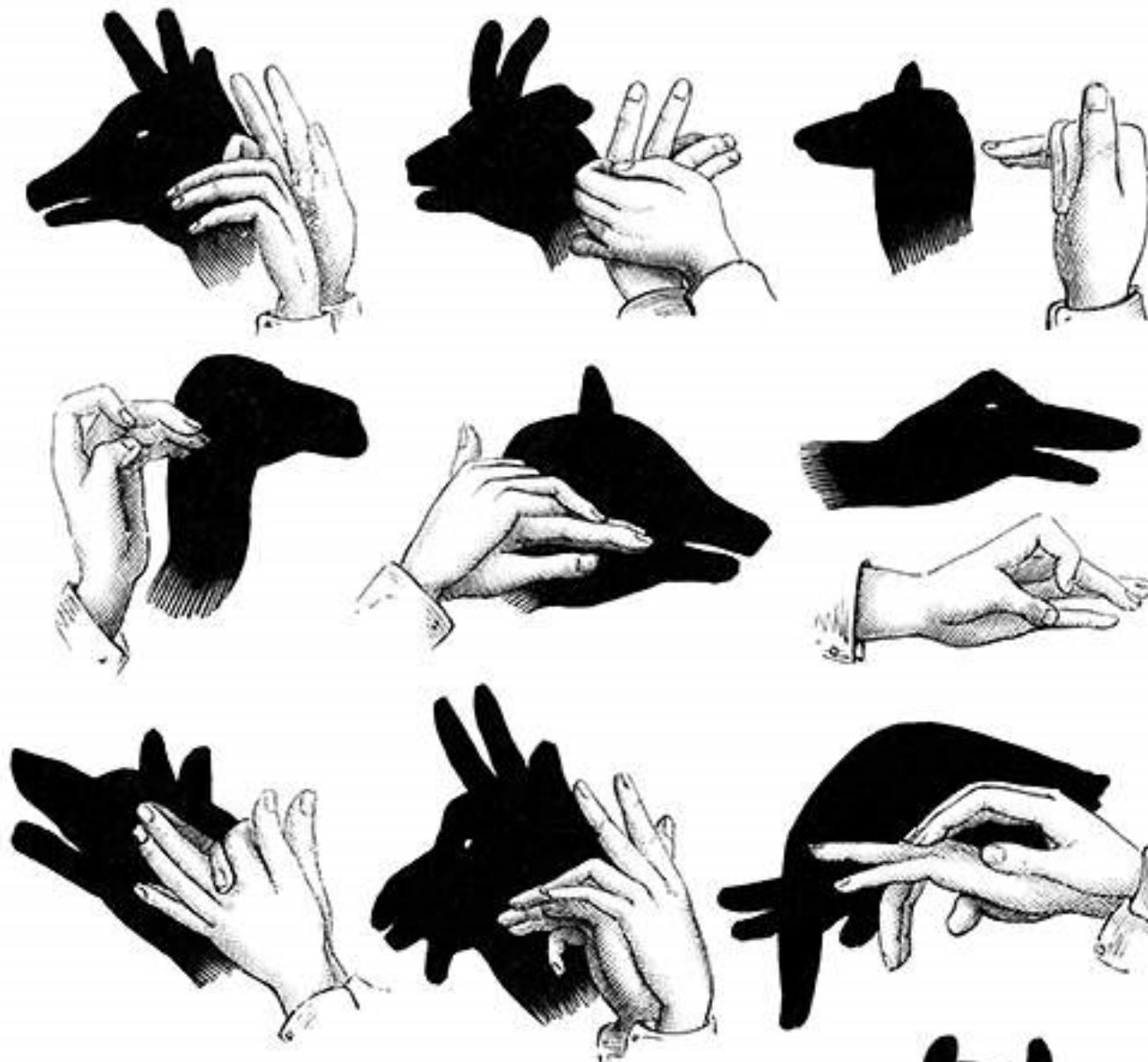
- In Section 6.2, we start with a collection of vectors instead of a collection of waveforms.
 - The calculation for vector case is easier.
 - It may seem redundant because some implicit axes are already there when we specify any vector.
 - Still useful because the calculation can remove the redundant axes.
 - The formula is exactly the same for the waveform case.
- In Section 6.3, the waveform case is studied.



Projection



Projection: Hand Shadow



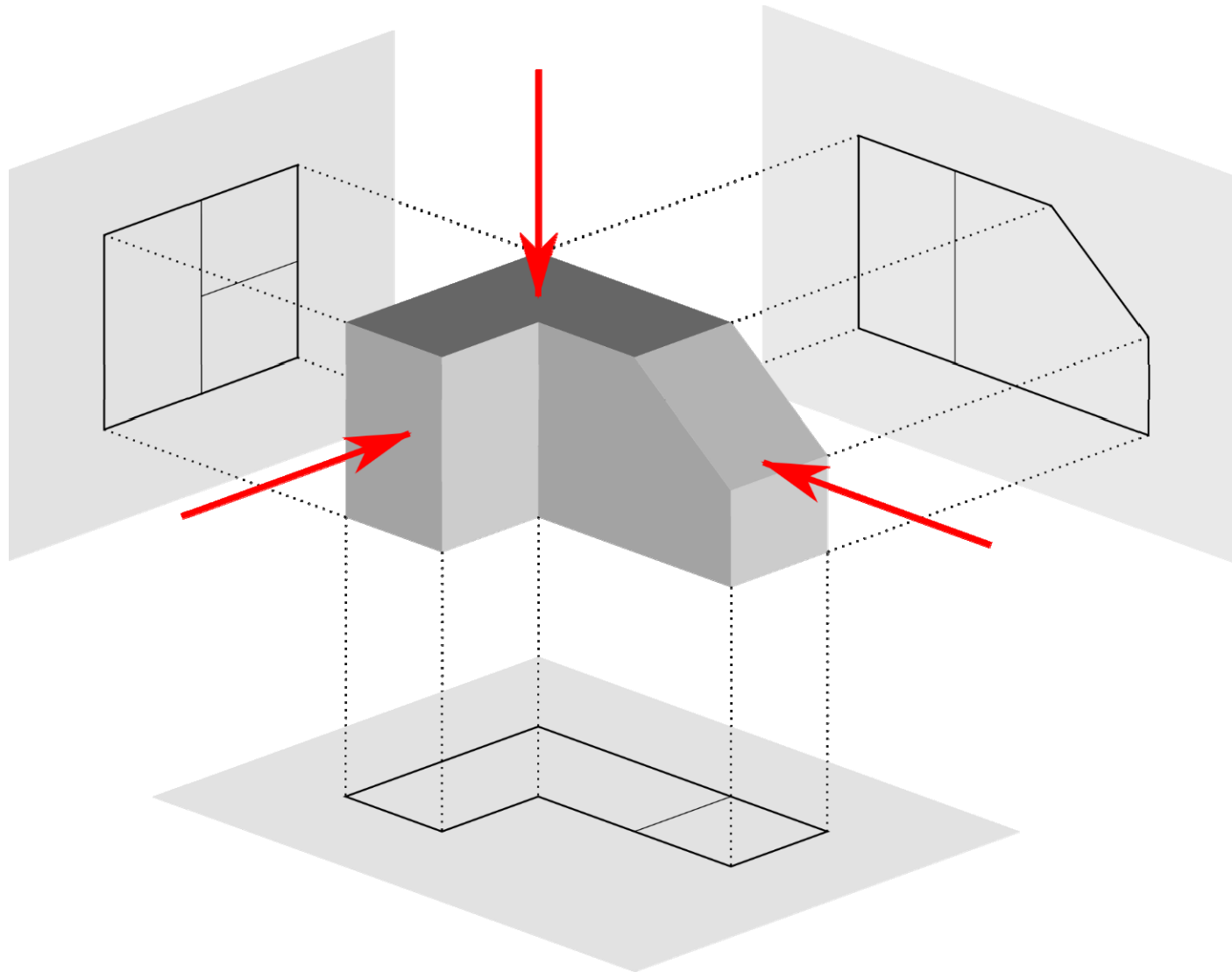
Projection: Shadow Art



Projection: Shadow Art



Projection: Engineering drawing

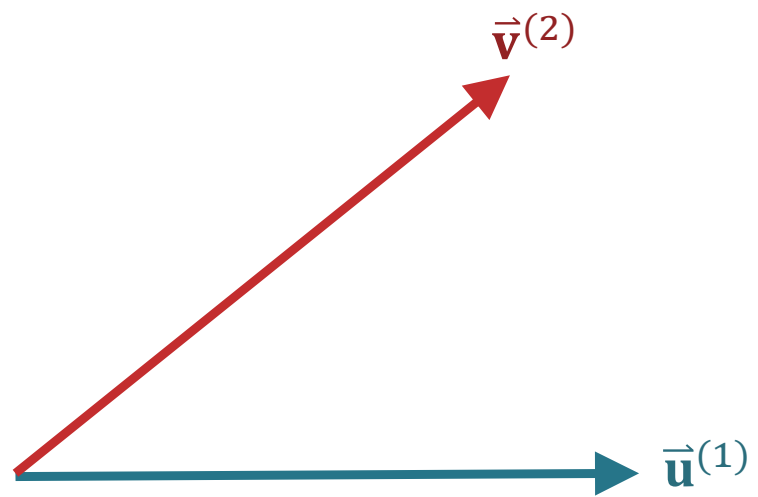


GSOP



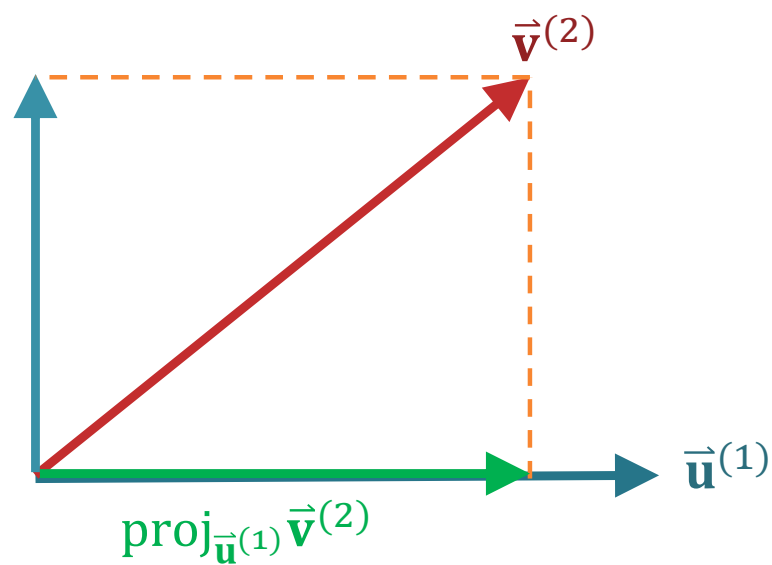
A diagram illustrating the relationship between vectors. A horizontal line is shown with a yellow arrow pointing to the right, labeled $\vec{e}^{(1)}$ below it. This yellow arrow is followed by a red arrow pointing to the right, labeled $\vec{u}^{(1)} = \vec{v}^{(1)}$ to its right. The red arrow is longer than the yellow arrow, and the two arrows are aligned along the same horizontal line, indicating that $\vec{u}^{(1)}$ and $\vec{v}^{(1)}$ are scalar multiples of $\vec{e}^{(1)}$.

GSOP



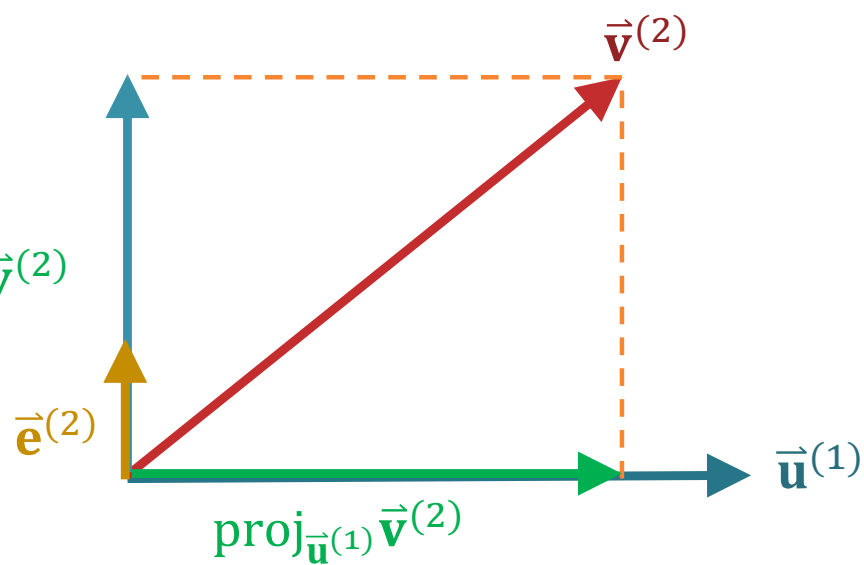
GSOP

$$\vec{u}^{(2)} = \vec{v}^{(2)} - \text{proj}_{\vec{u}^{(1)}} \vec{v}^{(2)}$$

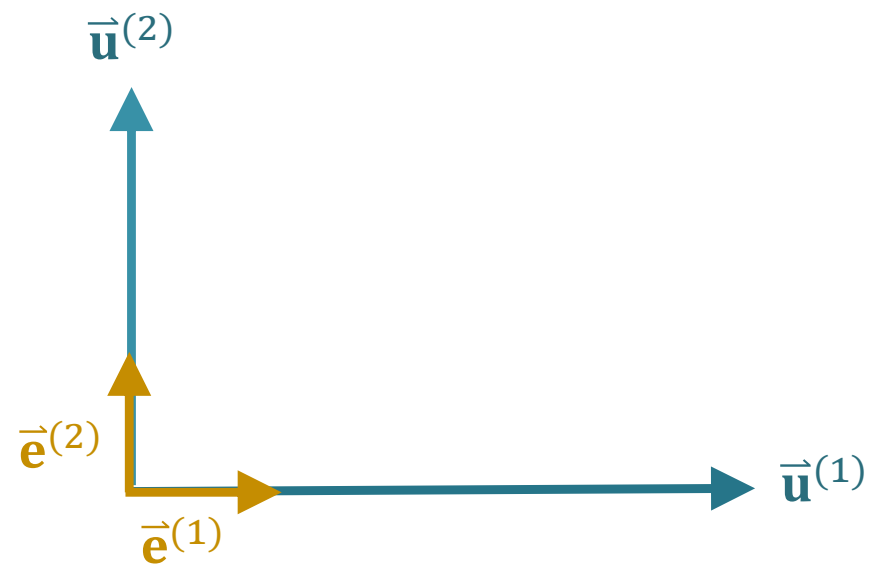


GSOP

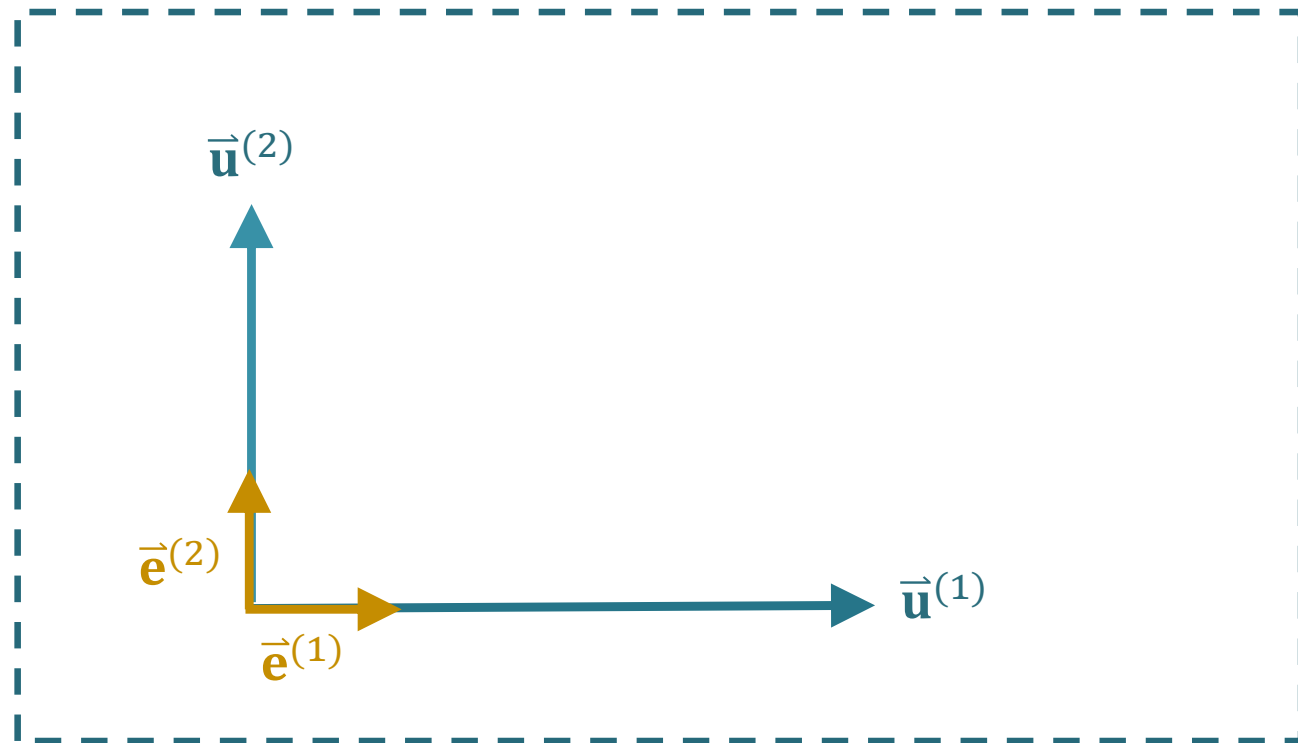
$$\vec{u}^{(2)} = \vec{v}^{(2)} - \text{proj}_{\vec{u}^{(1)}} \vec{v}^{(2)}$$



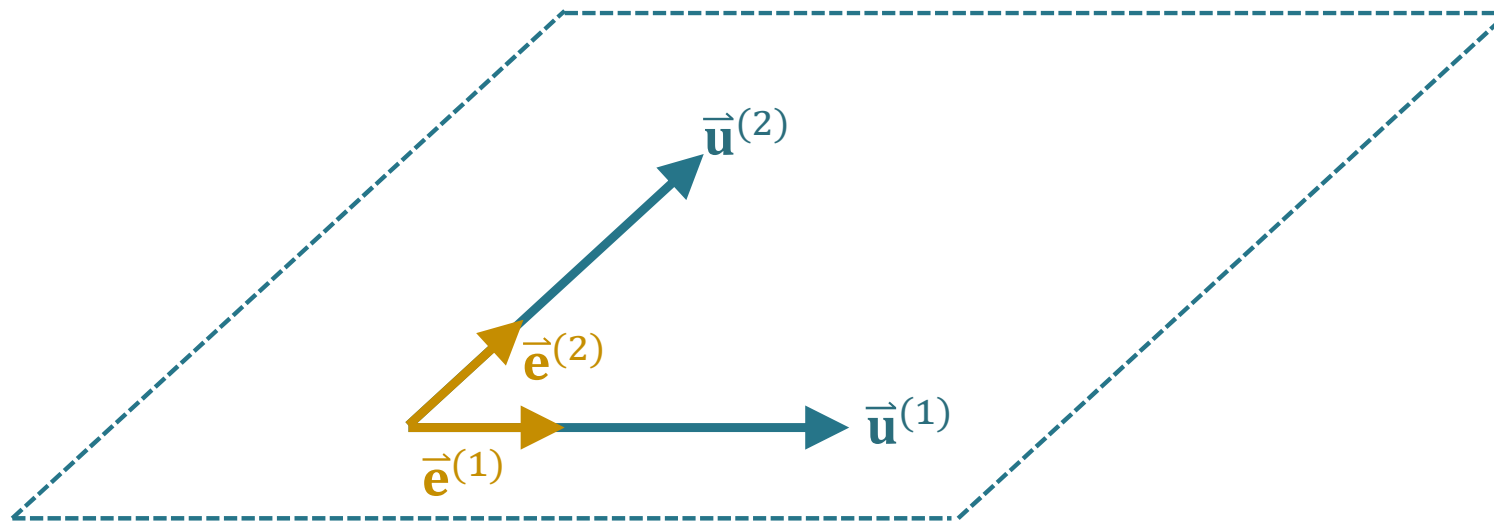
GSOP



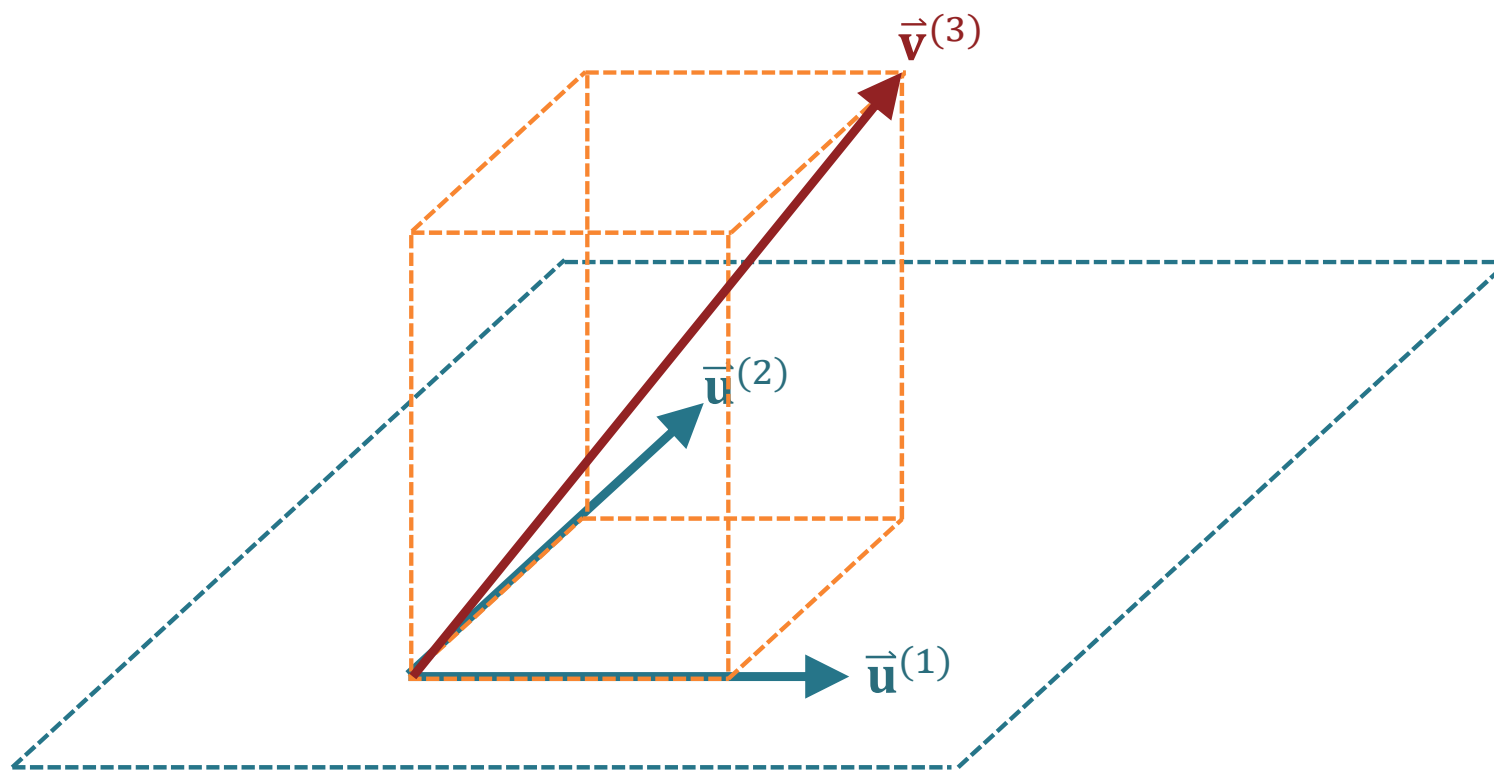
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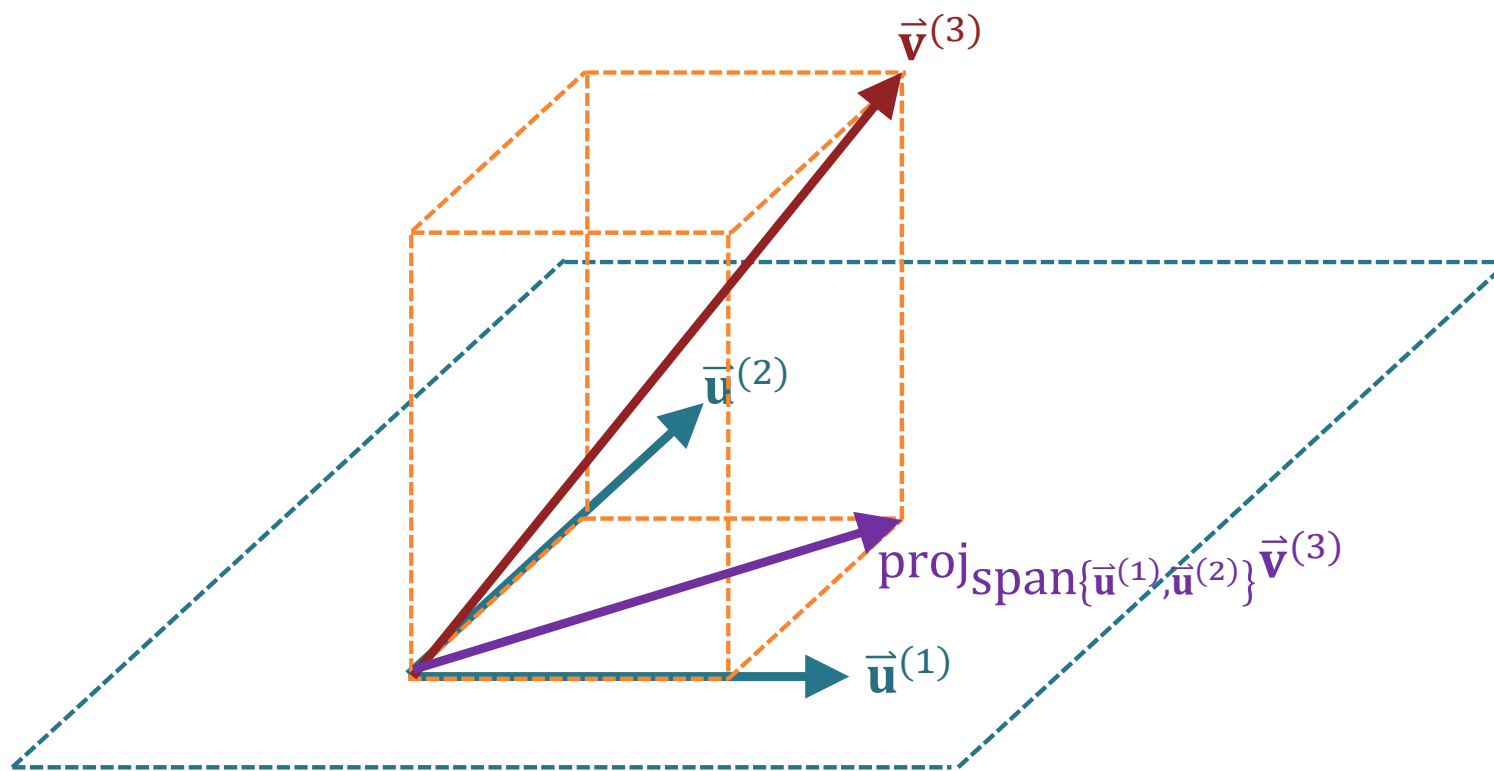
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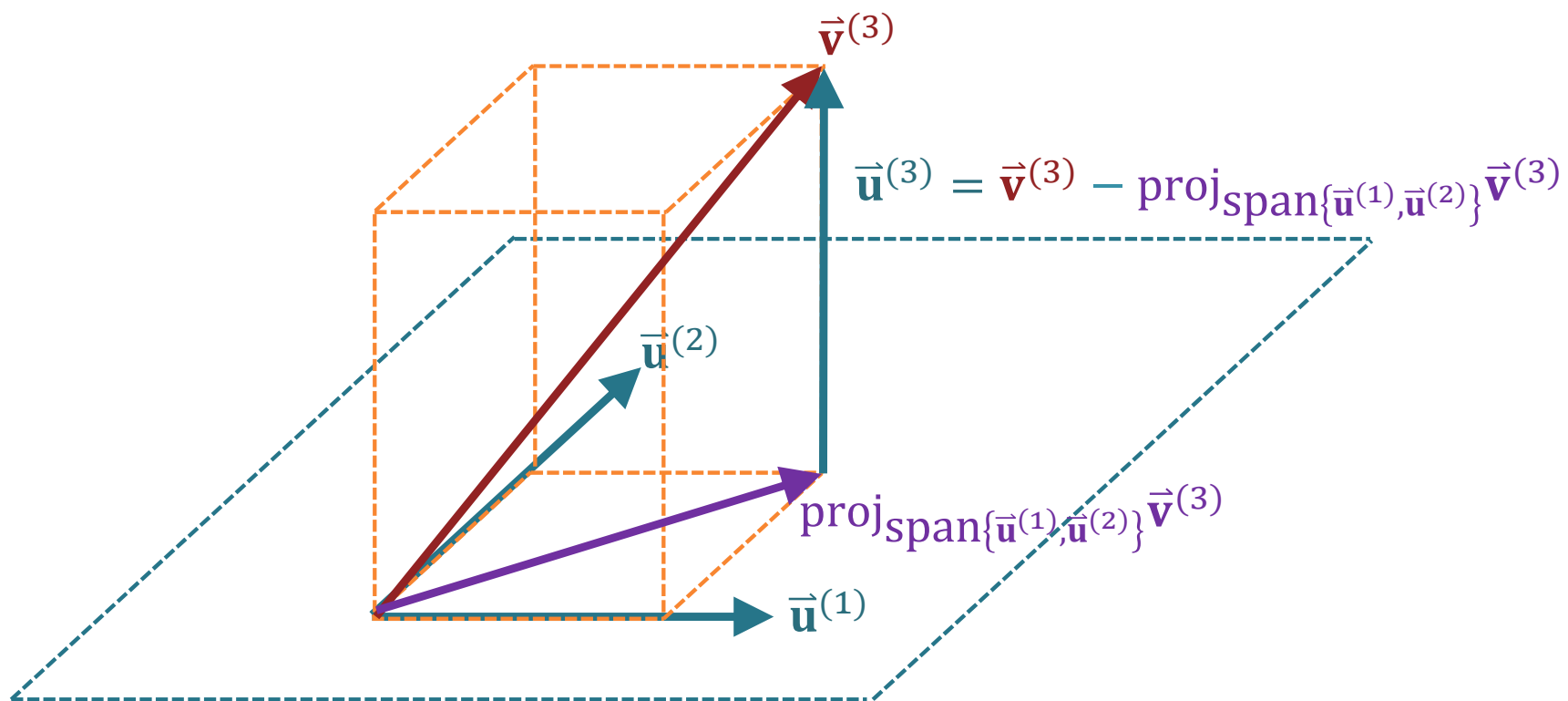
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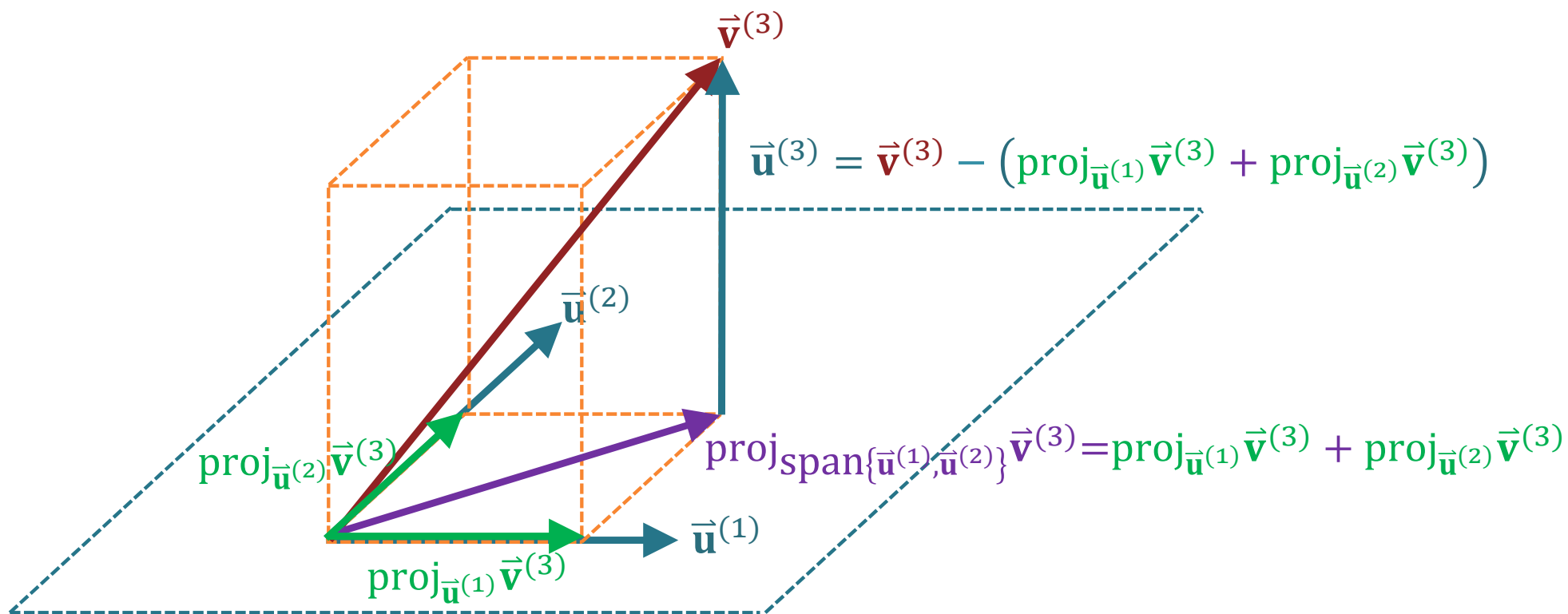
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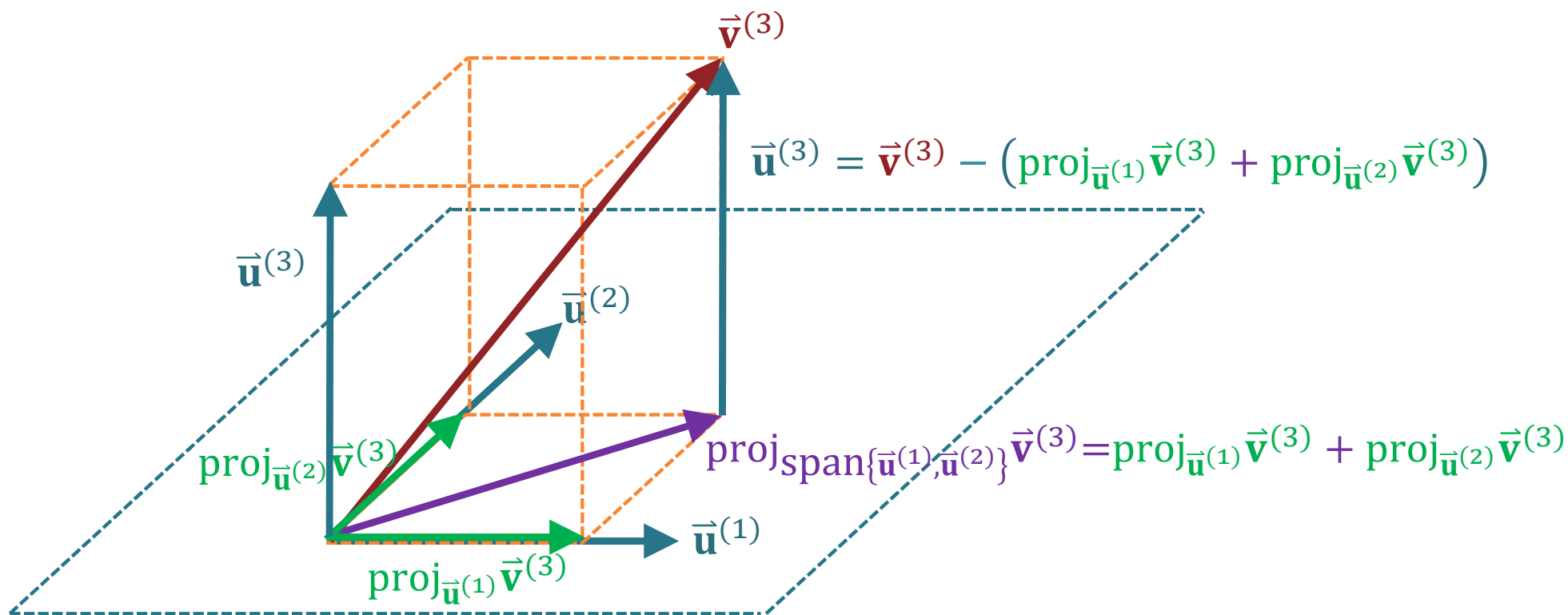
GSOP



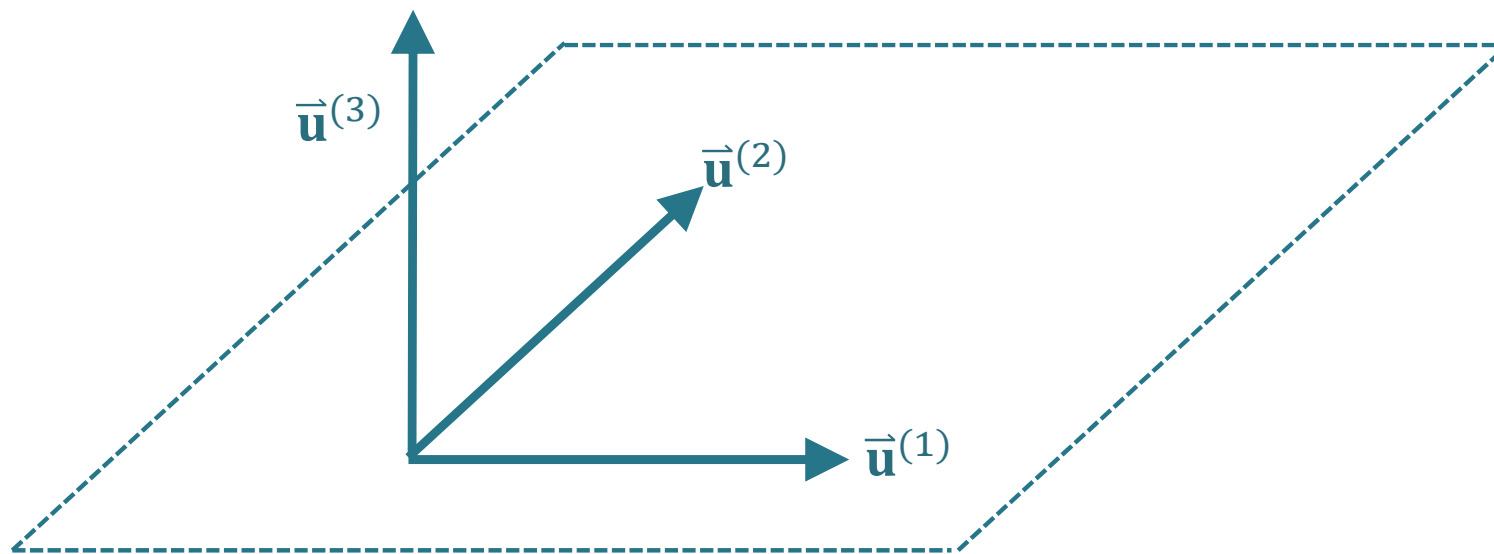
GSOP



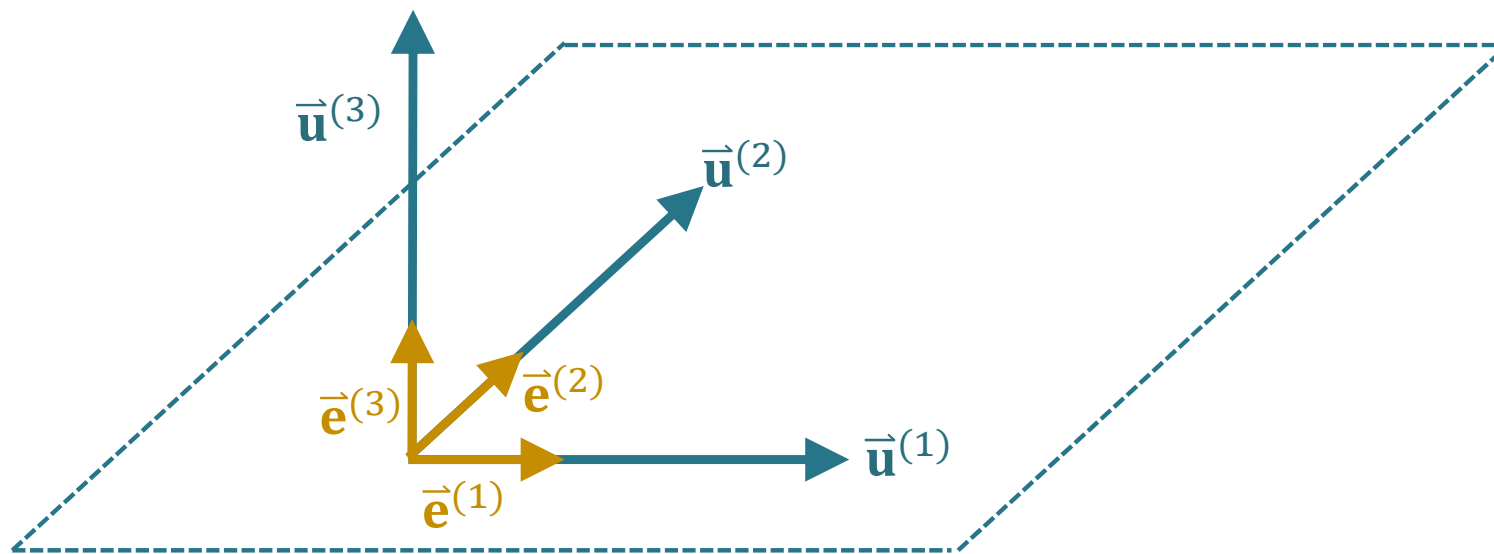
GSOP



GSOP

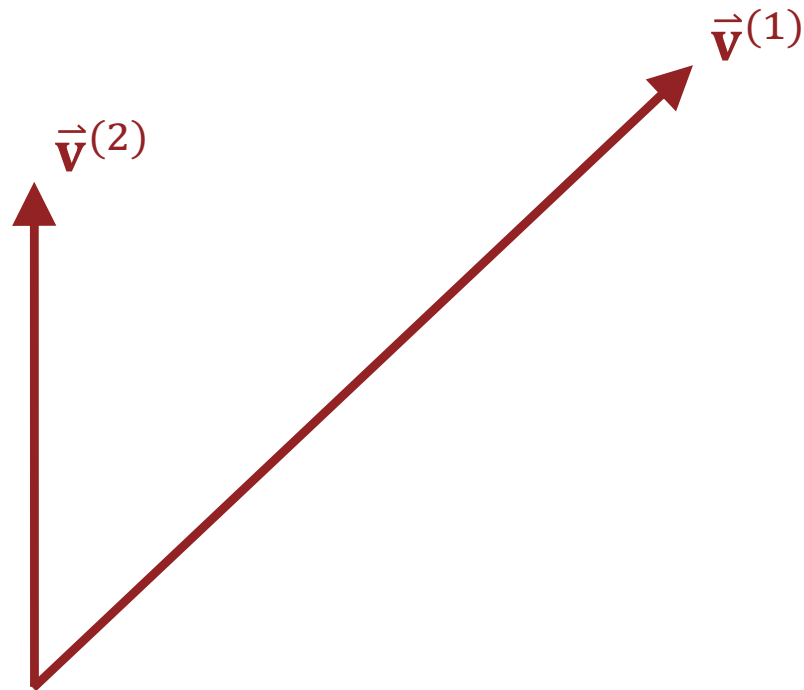


GSOP



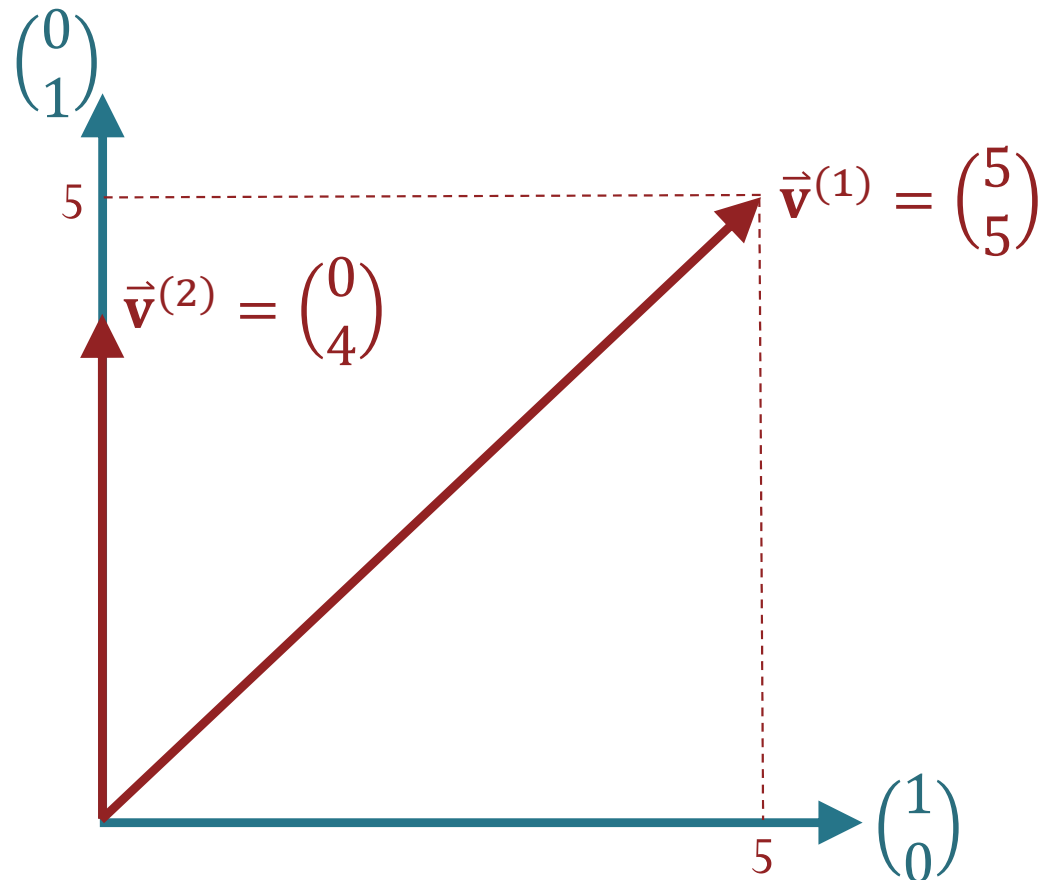
Changing basis / coordinate systems

Two vectors: $\vec{v}^{(1)}$ and $\vec{v}^{(2)}$



Changing basis / coordinate systems

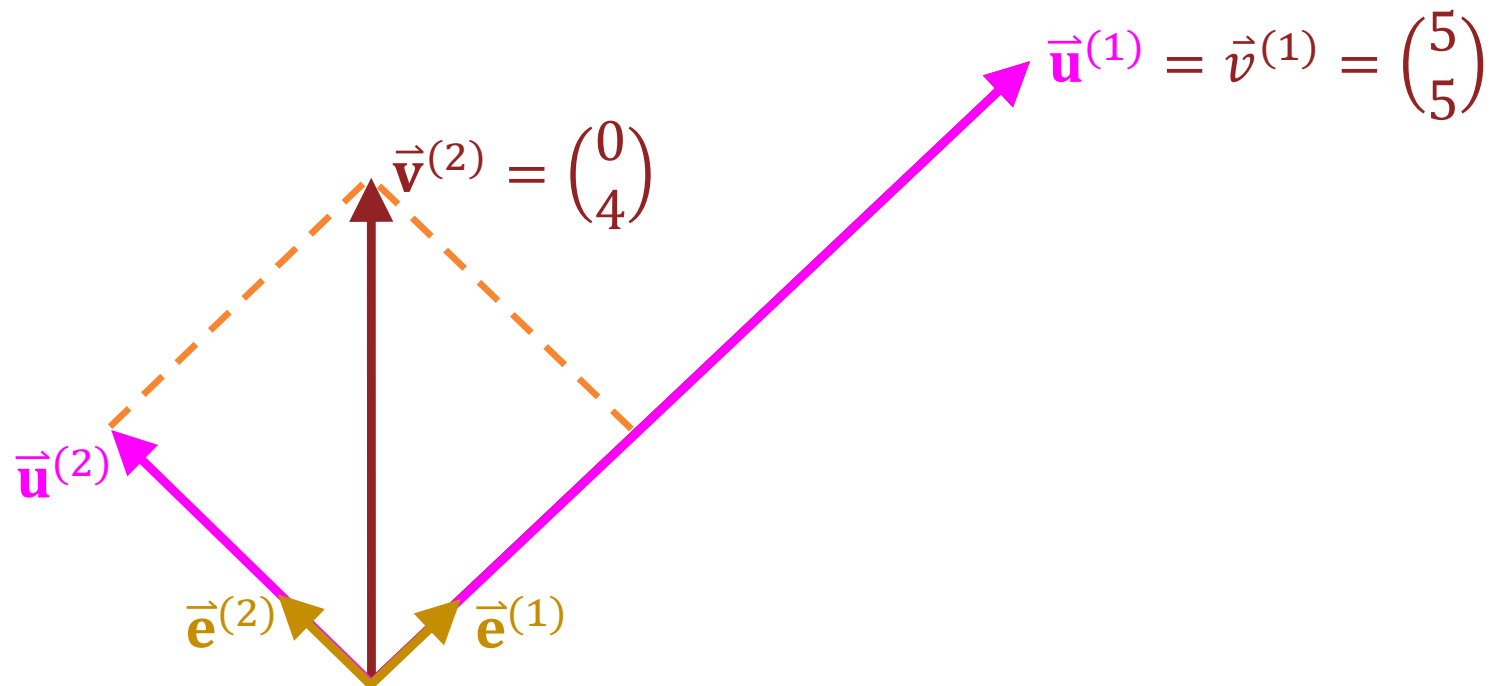
Old Coordinates: $\vec{v}^{(1)} = \begin{pmatrix} 5 \\ 5 \end{pmatrix}$ and $\vec{v}^{(2)} = \begin{pmatrix} 0 \\ 4 \end{pmatrix}$



Changing basis / coordinate systems

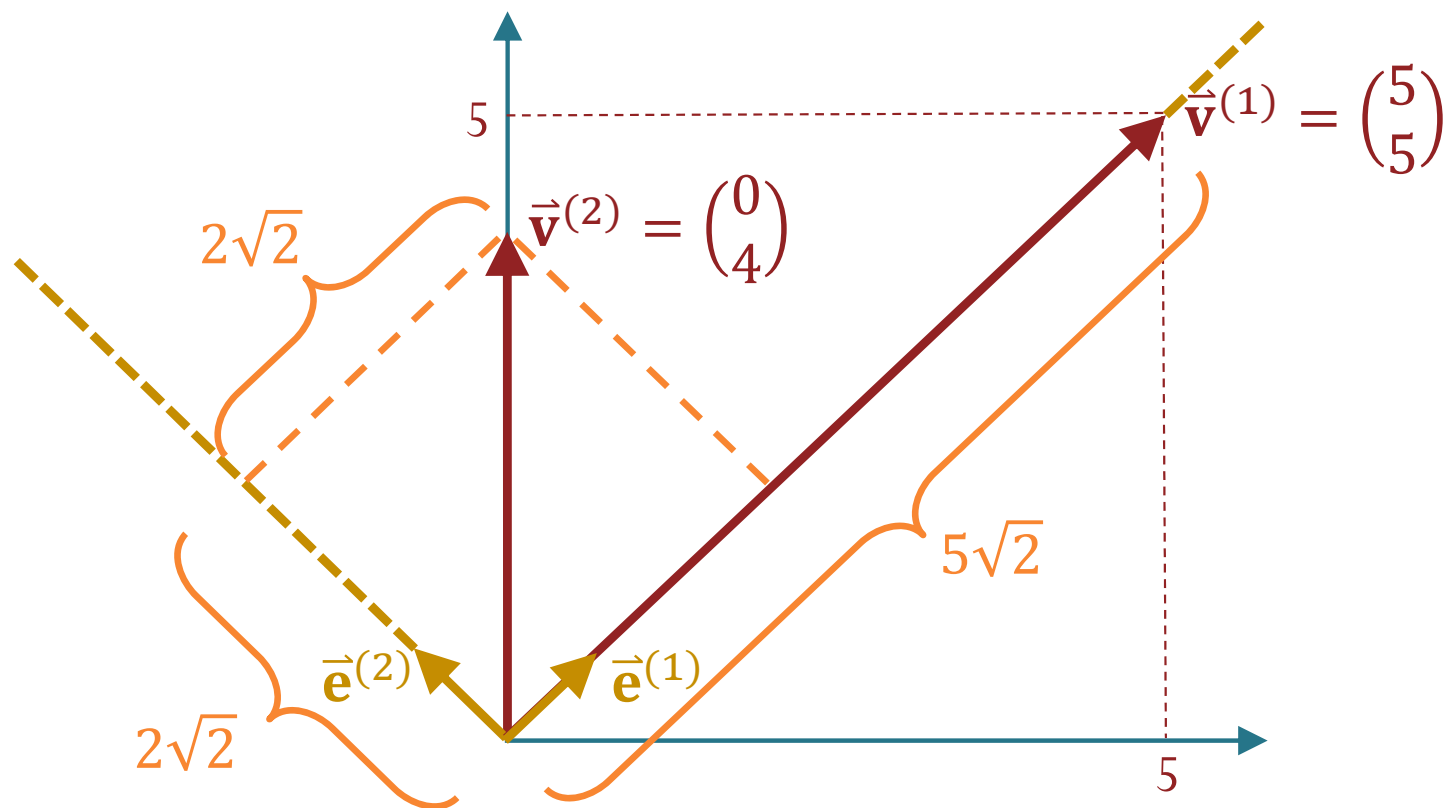
Old Coordinates: $\vec{v}^{(1)} = \begin{pmatrix} 5 \\ 5 \end{pmatrix}$ and $\vec{v}^{(2)} = \begin{pmatrix} 0 \\ 4 \end{pmatrix}$

Applying GSOP:



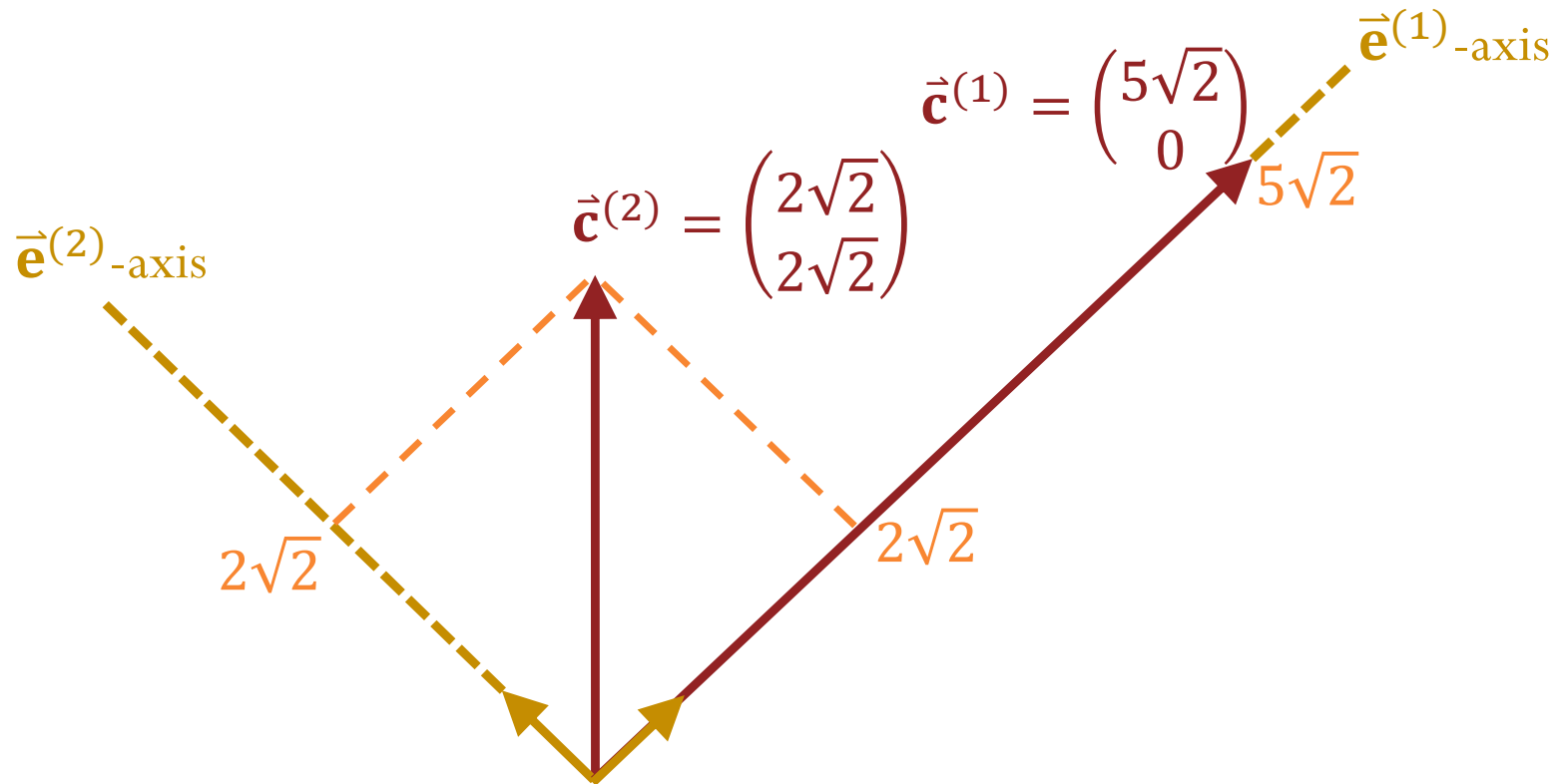
Changing basis / coordinate systems

New axes: $\vec{e}^{(1)}$ and $\vec{e}^{(2)}$

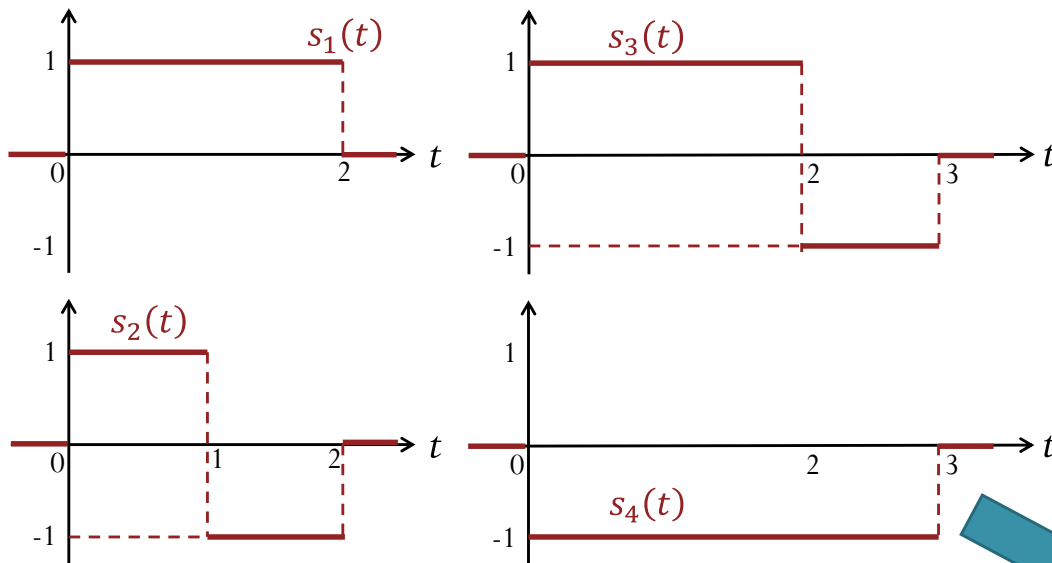


Changing basis / coordinate systems

New Coordinates: $\vec{c}^{(1)} = \begin{pmatrix} 5\sqrt{2} \\ 0 \end{pmatrix}$ and $\vec{c}^{(2)} = \begin{pmatrix} 2\sqrt{2} \\ 2\sqrt{2} \end{pmatrix}$



From Waveforms to Constellation

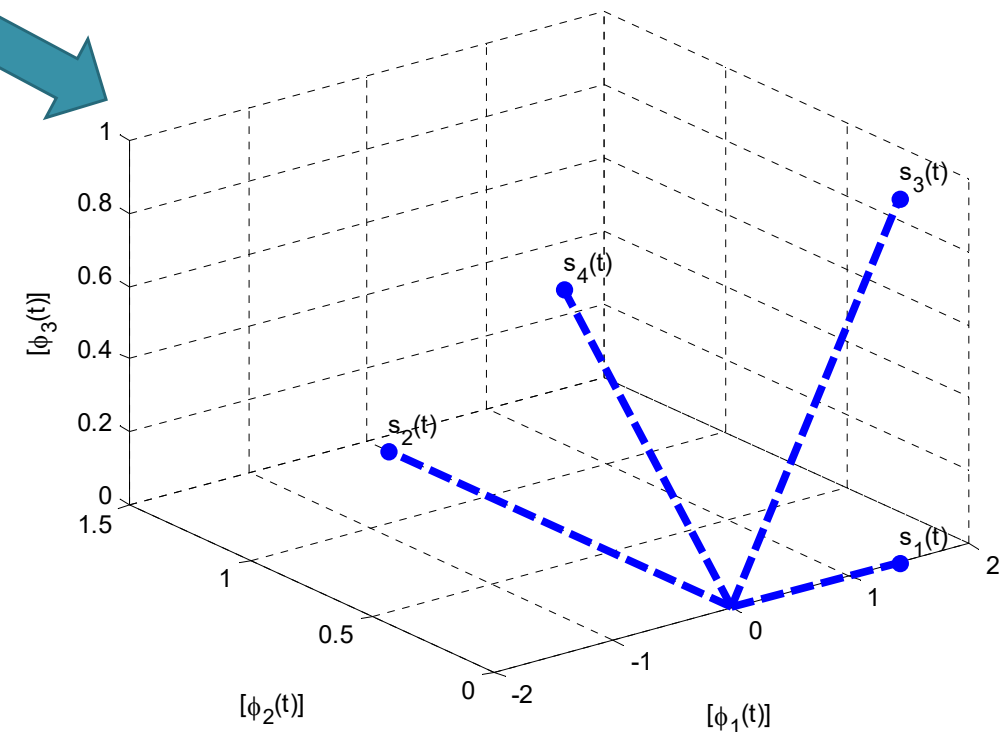


Use GSOP to find N orthonormal basis functions $\{\phi_1(t), \phi_2(t), \dots, \phi_K(t)\}$ to use as axes for $\{s_1(t), s_2(t), \dots, s_M(t)\}$.

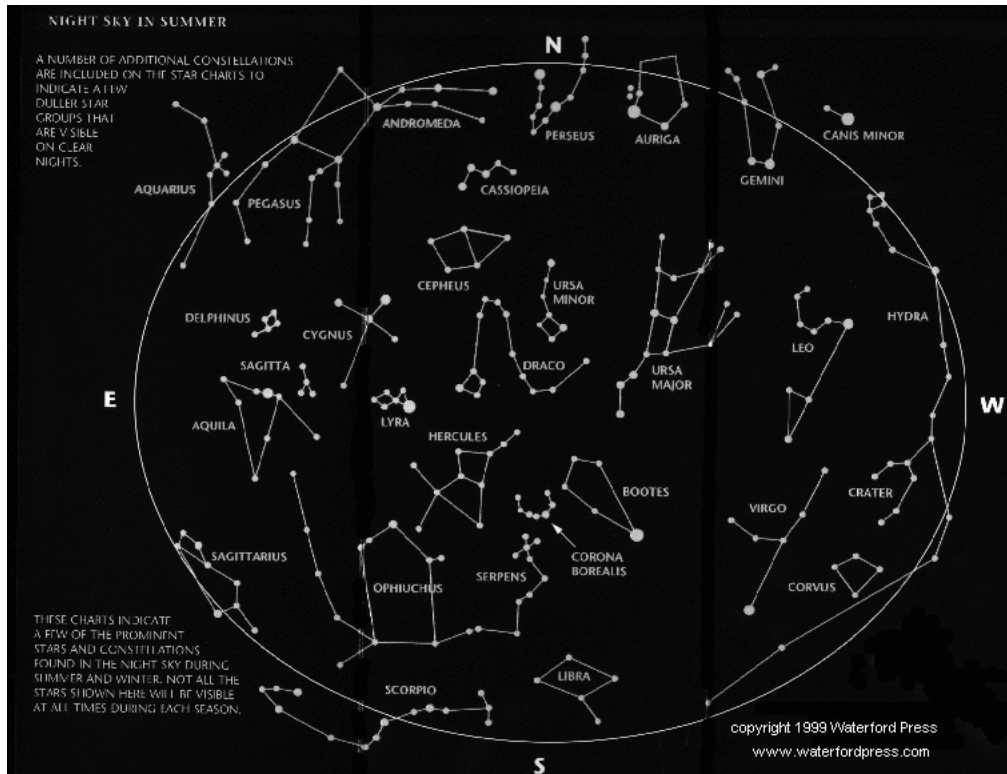
This gives vector representations for the waveforms $s_1(t), s_2(t), \dots, s_M(t)$:

$$\vec{s}^{(1)}, \vec{s}^{(2)}, \dots, \vec{s}^{(M)}$$

which can be visualized in the form of signal constellation



Star Constellations



[<http://iamintellectuallypromiscuous.com/science/star-right-straight-morning/>]



[http://68.media.tumblr.com/89eed4669ec511bbb6413acb8da8b0e2/tumblr_mz3qmyQeLH1rhh9f5o1_r1_400.jpg]

Review: QAM

4.8 Quadrature Amplitude Modulation (QAM)

Definition 4.82. In *quadrature amplitude modulation (QAM)* or *quadrature multiplexing*, two baseband real-valued signals $m_1(t)$ and $m_2(t)$ are transmitted simultaneously via the corresponding QAM signal:

Form *1 $x_{\text{QAM}}(t) = m_1(t) \sqrt{2} \cos(2\pi f_c t) + m_2(t) \sqrt{2} \sin(2\pi f_c t)$.

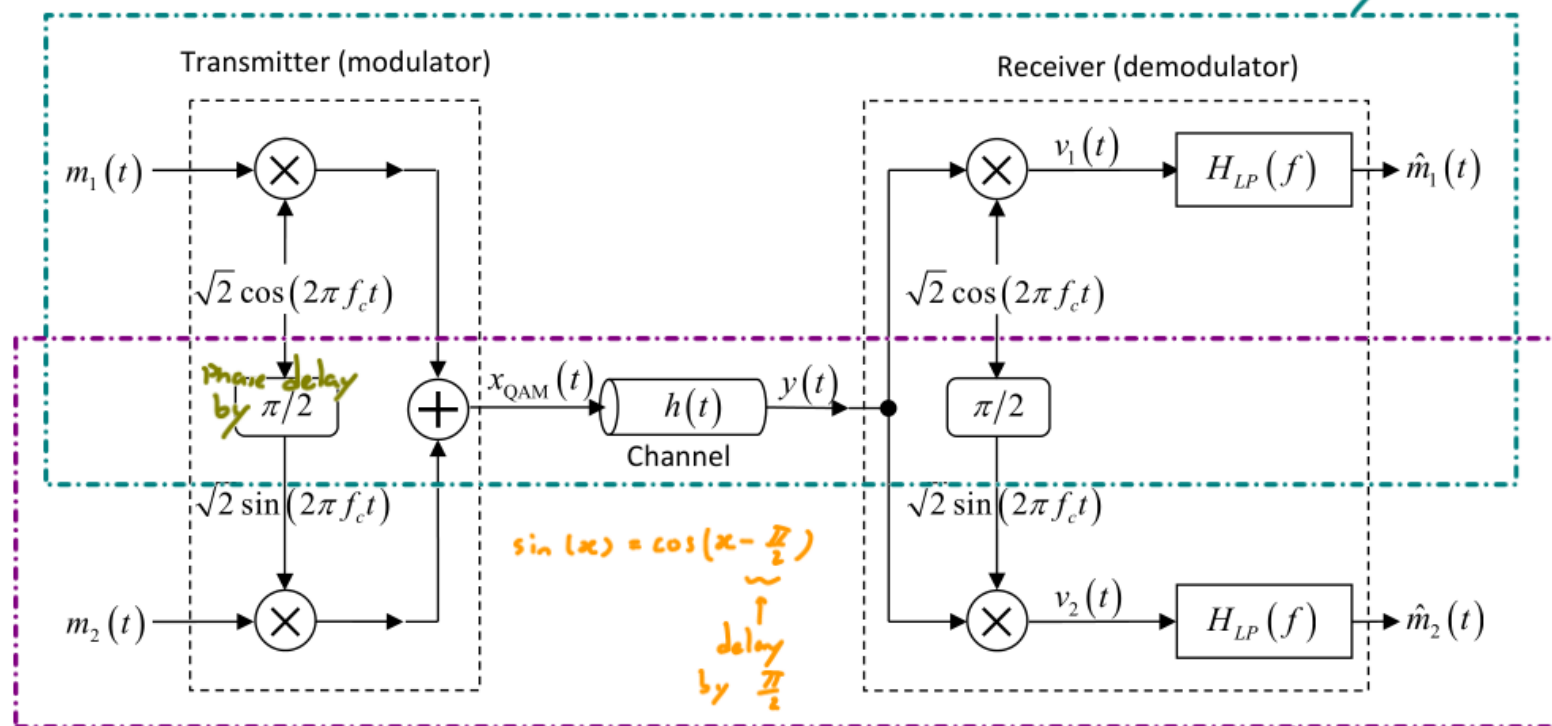


Figure 30: QAM Scheme

Review: QAM

4.87. Sinusoidal form (envelope-and-phase description [3], p. 165):

Form ✖ 2

$$x_{\text{QAM}}(t) = \sqrt{2}E(t) \cos(2\pi f_c t + \phi(t)),$$

calculator

where

envelope: $E(t) = |m_1(t) - jm_2(t)| = \sqrt{m_1^2(t) + m_2^2(t)}$

phase: $\phi(t) = \angle(m_1(t) - jm_2(t))$

4.90. Complex form:

$$x_{\text{QAM}}(t) = \sqrt{2}\text{Re} \{ (m(t)) e^{j2\pi f_c t} \}$$

where²⁰ $m(t) = m_1(t) - jm_2(t)$.

- We refer to $m(t)$ as the **complex envelope** (or **complex baseband signal**) and the signals $m_1(t)$ and $m_2(t)$ are known as the **in-phase** and **quadrature(-phase)** components of $x_{\text{QAM}}(t)$.



Standard Quaternary QAM

